ARRAY FACTOR OPTIMIZATION OF AN ACTIVE PLANAR PHASED ARRAY USING EVOLUTIONARY ALGORITHM

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ABSTRACT

Evolutionary algorithms (EAs) have the potential to handle complex, multi-dimensional optimization problems in the field of phased array. Out of different EAs, particle swarm optimization (PSO) is a popular choice. In a phased array, antenna element failure is a common phenomenon and this leads to degradation of the array factor (AF) pattern, primarily in terms of increased side lobe levels (SLLs), displacement of nulls and reduction in the null depths. The recovery of a degraded pattern using a cost and time-effective approach is on demand. In this context, an attempt made to obtain an optimized AF pattern after fault in a 49 elements quasi-circular aperture equilateral triangular grid active planar phased array using PSO. In the paper, multiple cases on recovery are discussed having a maximum 20% element failure. Each recovery is also further evaluated by different statistical analyses. A dedicated software tool was developed to carry out the work presented in this paper.

KEYWORDS

Active planar phased array, array factor (AF), particle swarm optimization (PSO), radar, transmit receive module (TRM)

1. INTRODUCTION

With rapid development and advancement in the technologies, the trend of incorporating an 'active' phased array configuration in majority of wireless, mobile, space and radar communication system is increasing day by day [1, 2]. Studies revealed that an active phased array configuration offers the collective advantages of high reliability, improved efficiency, low loss and higher sensitivity over a passive configuration [3]. An active phased array system contains several antenna elements assembled with individual solid state transmit receive module (TRM) to meet the required power aperture product. The system uses the abilities of active array to electronically steer the array beam instantaneously in any direction by controlling in-build phase shifters of TRMs [4].

Failure of single or multiple TRMs in such systems is a common occurrence which makes an antenna element become non-radiative. In the event of the failure, the far field array factor (AF) pattern of the system degrades seriously in terms of increased side lobe levels (SLLs), displacement of nulls and reduction in the null depths. To get rid of such degradations, repair/ replacement of the faulty modules could be best choice, as it always gives the best recovery of the original AF pattern. However, such repair/ replacement in a space based or ground based systems commissioned with hundreds of TRMs will be expensive in terms of time and budget. Under such circumstances, without undergoing any repair/ replacement, evolutionary algorithms (EAs) could

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be used as an alternative to recover the optimized pattern closest to the original by perturbing the input amplitudes and/ or phases of input excitations of remaining functional TRMs.

EAs are inspired by the biological models of evolution and natural selection [5, 6]. To solve a particular problem using EAs, a situation is created in which potential solutions can evolve. The situation is outlined by the parameters of the problem that pushes the algorithm towards optimal solutions. Most commonly used EAs are genetic algorithm (GA) [7], particle swarm optimization (PSO) [8, 9], cuckoo search (CS) [10], and artificial bee colony (ABC) [11]. Several studies have been reported on optimization of AF pattern by amplitude and/ or phase perturbation using EAs for linear array [12 - 15]. The present work deals with recovery of the AF pattern of an active planar phased array by recalculating a new set of amplitudes of input excitations of working elements using PSO. Being a stochastic technique and based on the movement and intelligence of swarms, PSO has gained an immense popularity in the field of electromagnetics [16]. The reason behind selecting PSO is that PSO algorithm has the advantages of its simplicity and less computational complexity over the other EAs. The computational time taken by PSO to arrive at the desired solution is also less as compared to other EAs [17 - 19].

This paper presents multiple cases on the recovery of the AF pattern of a 49 elements active planar phased array operating in very high frequency (VHF) band. Active array in VHF band has wide application in the areas of wind profiling and weather radar [20]. In the presented cases, the number of non-working elements is increased to a maximum 20% of the total elements. The recovered pattern in each case is evaluated numerically by comparing the levels of first side lobes (FSLs) and the depth of the first nulls (FNs) with the corresponding values in the original AF pattern.

2. ARRAY CONCEPT

The AF pattern of an active planar phased array consisting of $M \ge N$ identical elements placed on XY plane (Fig. 1) scanning to a direction (θ_o, φ_o) may be expressed as [4]:

$$AF(\theta, \varphi) = \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} \left| A_{mn} \right| e^{j[m(T_x - T_{xs}) + n(T_y - T_{ys})]}$$
(1)

where, d_x and d_y = inter-element spacing along X and Y direction respectively; $T_x = (2\pi/\lambda) d_x \cos(\alpha_x)$; $T_y = (2\pi/\lambda) d_y \cos(\alpha_y)$; $T_{xs} = (2\pi/\lambda) d_x \cos(\alpha_{xs})$; $T_{ys} = (2\pi/\lambda) d_y \cos(\alpha_{ys})$; $\cos(\alpha_x) = \sin(\theta)\cos(\varphi)$; $\cos(\alpha_y) = \sin(\theta)\sin(\varphi)$; $\cos(\alpha_{xs}) = \sin(\theta_o)\cos(\varphi_o)$; $\cos(\alpha_{ys}) = \sin(\theta_o)\sin(\varphi_o)$; and $A_{mn} =$ Amplitude of input excitation of the mn^{th} element.



Fig. 1. A planar phased array

The phasing of mn^{th} element, ψ_{mn} for beam steering is calculated as $mT_{xs} + nT_{ys}$ and the far field radiation pattern of the array is:

$$ff(\theta, \varphi) = AF(\theta, \varphi) \times EP(\theta, \varphi)$$
⁽²⁾

where, $EP(\theta, \varphi)$ is the Element Pattern (EP) of an antenna element at (θ, φ) plane.

3. PSO

EAs are stochastic optimization techniques popularly being used for real life problem solving tasks in wide and diverse range of application areas like science, engineering, medical and economics [5, 6] and [21]. Among all EAs inspired from biological process, the PSO technique is simple, effective and capable of handling tricky multidimensional nonlinear optimization problems across the fields [16 - 19]. The PSO was originally introduced in 1995 by Russell Eberhart, an electrical engineer, and James Kennedy, a social psychologist [8, 9]. Here it is necessary to mention that the present work is not intended to carry an extensive review on PSO, and therefore only the main steps of the algorithm are discussed.

The PSO is a population based search strategy which is applied in an N dimensional search space containing set of points, referred as *particles*. These set of particles moves in the search space with velocities that are dynamically adjusted according to their historical performance as well as the experience and knowledge of its neighbours. During movement, at each iteration, the particles update their velocities according to the best positions already found by themselves as personnel best (P_{best}) and global best (G_{best}) [16].

$$v_{n+1} = w.v_n + c_1.rand_1 \left[\frac{P_{best, n} - X_n}{\Delta t} \right] + c_2.rand_2 \left[\frac{G_{best, n} - X_n}{\Delta t} \right]$$

inertia
cognitive(personal)component isocial(neighbourhood)component (3)

where, v_{n+1} and v_n are the velocity vectors of the particle in the present and previous iterations respectively; X_n is the current position of the particle; w is the inertial weight which varies as the algorithm progresses and determines the weight by which the particles current velocity depends on its previous velocity; c_1 is the cognitive coefficient controlling the pull to the P_{best} position; c_2 is the social rate coefficient controlling the pull to the G_{best} position; $rand_1$ and $rand_2$ are the random numbers uniformly distributed in (0, 1); Δt is the time difference between the two successive positions of the particles and is most often taken as unity. The next position, X_{n+1} of particle in the search space is calculated as:

$$X_{n+1} = X_n + v_{n+1} \Delta t \tag{4}$$

The choice of different parameters and the constants in above two equations can be decided according to the work described in [22–25]. During the process of algorithm, the P_{best} and G_{best} are decided according to the criteria set by a fitness function (FF) [16]. A FF plays a vital role in controlling the performance of the algorithm. The selection of FF depends on the type and complexities present in the identified problem.

4. METHODOLOGY OF THE WORK

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In the present work, a planar phased array having a quasi-circular aperture is considered as the test array. The diagram of the array shown in Fig. 2 is populated with 49 elements, each represented by a square box labeled with a fixed numerical number for easy identification of element location. Elements of the array are arranged as equilateral triangular grid in 9 rows and 15 columns. During the design of the array geometry, it is attempted to make the periphery of the aperture close to circular to maintain angular symmetry in the AF pattern. A configuration of a planar array having circular or near to circular aperture with triangular grid gives a grating lobe free AF pattern at maximum possible off-zenith angle with comparatively low FSLs without any amplitude tapering. Triangular grid also helps to reduce the number of elements demanded to fill a certain aperture [26]. These inherent advantages make the configuration a popular choice in applications like wireless, space borne synthetic aperture radar (SAR), weather radar, wind profilers [20], [27-29].



Fig. 2. The test array

EP of the array elements is considered as isotropic and initial amplitude of input excitations are taken as unity. The inter-element spacing is kept 0.7λ which places the rows and columns at 0.88 m and 1.02 m apart, respectively. Regardless of aperture or grid structure, the AF equation described in equation (1) takes individual element location as inputs. The same equation is used for calculating the AF of the test array. Effect of mutual coupling among elements is not considered during the analysis and discussion.

After finalizing the structure of the test array, a suitable FF is developed and expressed as equation (5). There could be many possibilities of the FF depending on the factors decided to achieve the optimization. The present FF is developed after a number of trials to bring the solution close to the optima by minimizing it till last iteration. The weights (a_1 , a_2 , a_3 and a_4) in the FF equation are taken as unity.

$$FF = a_1F_1 + a_2F_2 + a_3F_3 + a_4F_4$$
(5)

where,

,

$$\begin{aligned}
F_2 &= [AF_{FSL_pso} - AF_{FSL_original}]^2 \\
F_3 &= [AF_{FN_pso} - AF_{FN_orig}]^2 \\
F_4 &= \sum_{\theta=-90^\circ}^{+90^\circ} [AF_{pso}(\theta, \theta) - AF_{original}(\theta, \theta)]^2
\end{aligned}$$

 $F_1 = [AF_{pso}(0,0) - AF_{original}(0,0)]^2$

Balanis says that the characteristic of an AF pattern can be broadly determined by the three levels: main beam, FSLs and FNs [30]. F_1 , F_2 and F_3 of FF uses these three levels of optimized and original patterns for deciding movement of particles in problem space. To test the

effectiveness of the PSO algorithm for the present problem with the FF, a software tool is developed. The coding of the tool is done based on the array concepts and the PSO algorithm. The tool takes the values of all the PSO parameters and the location of non-working element as user inputs. Research revealed that PSO can achieve faster convergence towards optima with less oscillation by careful selection of its parameters [22 – 25]. After a detailed study, following values of the PSO parameters are fed into the developed tool and kept fixed in all the presented cases.

No of particles = 200, No of iterations =100, Cognitive component $(c_1) = 2$, and Social component $(c_2) = 1$

Different schemes on variation of 'w' of equation (3) are tested, and finally for faster convergence 'w' is linearly decreased over 100 iterations from 0.9 to 0.4. The same tool can be used for optimization of any configurations of planar phased array by modifying the elements co-ordinates.

In the code, A_{mn} of equation (1) is written as $(1 + \mu_{mn})$, where, μ_{mn} is introduced to represent the perturbation required in initial amplitude of the input excitations of working elements for recovery. Each iteration produces a new set of μ_{mn} which is algebraically added with unity to generate a new set of amplitudes for calculating the AF in present iteration required by all four entities of FF. The perturbations are restricted within a window of ±0.3, so the final amplitude factor, A_{mn} is restricted within a boundary between 0.7 and 1.3, which is equal to ± 30% perturbation of initial amplitude. The window in amplitude factor can be linked with the tuning of power amplifier modules in TRMs. In practical application, depending on the type and quality of a power amplifier module, the set window may vary. Based on the methodologies, the results obtained after single run for each case are presented and discussed in the next section.

5. RESULTS AND DISCUSSION

The original AF pattern of the test array in normalized form without any element failure at zenith $(\theta_{o_i} = 0^\circ, \varphi_o = 0^\circ)$ is depicted in Fig. 3. From the figure, it is noted that the levels of FSLs are - 19.08 dB down whereas the depths of FNs are -38.84 dB down from the peak level of the main beam which is at 0 dB after normalization. Throughout the work, the recovery of the pattern at zenith is only presented. No beam steering to other elevation (θ_o) or azimuth (φ_o) plane is introduced during the pattern optimization. After recovery in each case, the original AF, AF with fault and PSO optimized AF patterns are presented in a single figure for visual comparison and performance evaluation.



Fig. 3. Original AF pattern of the test array

In the next couple of paragraphs detailed evaluations on the AF pattern recovery in four different cases using the developed methodologies are discussed. Work starts with the recovery of the original pattern when a single element is made non-working. Thereafter, number of non-working elements are increased in subsequent cases to three, five and ten, respectively. The locations of the non-working elements are picked up randomly.

<u>*Case-1*</u>: In the first case the center element at location 1 is made non-working. The presence of a non-working element at the said location increases the levels of both FSLs and FNs by 1.64 dB and 7.16 dB, respectively. After creation of the fault, now the tool runs with set PSO parameters to recover the original pattern. At the end of 100th iteration with 200 particles, the tool produces an optimized AF pattern close to the original one shown in Fig. 4. The tool has recovered the pattern by generating a new set of amplitudes of input excitations of 48 working elements and brings down the levels of FSLs and FNs to -19.09 dB and -38.87 dB, respectively resulting very close matching with the corresponding values in the original pattern.



Fig. 4. AF patterns for case-1

<u>*Case-2:*</u> Work advances with three elements failure at locations 5, 2 and 28 selected randomly as user inputs. With these three non-working elements, the levels of FSLs increase to -16.52 dB. But in opposite of the case-1, the depths of FNs increase to -49.67 dB with one degree shift towards the main beam. Again the tool runs to recover the original pattern by amplitude perturbation of remaining 46 working elements. Both fault and PSO optimized patterns with original one are depicted in Fig. 5.



Fig. 5. AF pattern for case-2

The figure illustrates that the optimized pattern nicely follows the original pattern at every location of θ with hardly any variation in the levels. The shifting of FNs after faults is also rectified by placing them exactly at their original locations.

<u>Case-3</u>: Till now the tool has shown satisfactory performances in recovering the optimized pattern closest to the original after the occurrence of faults. But to imitate a more complex situation in an active phased array having hundreds of TRM, the tool is tested further with 10% non-working elements i.e. 5 of 49 elements. The location numbers of non-working elements are randomly taken as 37, 12, 30, 1 and 18. Remaining 44 working elements with a void at said locations produces a pattern with phenomena entirely opposite to the occurrences in the last two cases. FSLs of the fault pattern remain at very close to its original levels with one degree shift towards the main beam, whereas the second side lobes (SSLs) increase from -20.87 dB to -15.9 dB. These increased SSLs may pick up a strong interference signal from any external source and may raise the probability of occurrence of ambiguous detection [4] and [30]. Fig. 6 plots the AF pattern with faults at the said locations. The plot also depicts that the optimized pattern achieved after the final iteration is made highly resemble the original pattern. After optimization, the maximum levels of SLLs are brought down to -20.84 dB close to the original value -20.87 dB. In optimized pattern, the shifts in FSLs are rectified and original levels are also maintained.



Fig. 6. AF pattern for case-3

<u>*Case-4*</u>: To add much more complexity, the number of non-working elements is increased further to ten which is equal to 20 % of the total elements. The position numbers of the fault elements are again randomly selected as 49, 42, 13, 5, 10, 32, 31, 19, 26 and 17. Fig. 7 presents the AF pattern before and after faults. The fault pattern is deformed to a typical formation. The lobes after second nulls are merged together and pop up to maximum -15.98 dB at $\pm 30^{\circ}$. This merging eliminates the third and fourth nulls in both sides of the main beam. In opposite, on occurrence of merging, FSLs shift downward by approximately 10 dB. This downward shifting of FSLs could be better at the cost of the deformation, but the popping up of second and third side lobes could damage the entire application by inviting strong interferences into the system.





Fig. 7. AF pattern for case-4

Once again the tool runs in the quest of searching a new set of amplitudes for the remaining 39 working elements for correcting the failed pattern. The optimized pattern after recovery is presented in Fig. 7. The pattern illustrates that the process of recovery is successfully handled by the tool to a great extent. FSLs are pulled up to the values of the original pattern. The region after FSL in both sides of the main lobe is restructured at every point and brought near to the original values.

The convergence profiles of 200 particles obtained by minimizing the FF are presented in Fig. 8. The figure illustrates that the profiles of convergence are nicely damped and reached closest to optima at around 70^{th} iteration. Out of 200 particles, the 81^{th} particle emerges as the global best to produce the best optimized AF pattern.



Fig. 8. Convergence profiles of the FF for case-4

The final amplitudes of input excitations of 39 working elements of *case-4* are displayed in Fig. 9. The values of excitations are represented by a grey scale whose values vary between 0.6 and 1.3. The figure depicts the final excitations searched by the tool within the set window. In the figure, the non-working elements are represented by only number without any shaded box.



Fig. 9. Final amplitude distribution for case-4

A comparative summary on the values of the critical parameters of AF patterns in original, fault and recovered conditions discussed in above four cases is presented in Table-1.

Cases	Parameters	Original	After fault	After recovery	
Case-1	FSLs	-19.08 dB	-17.44 dB	-19.09 dB	
	FNs	-38.84 dB	-31.68 dB	-38.87 dB	
	FNBW	28°	26 °	28 °	
Case-2	FSLs	-19.08 dB	-16.52 dB	-19.08 dB	
	FNs	-38.84 dB	-49.67 dB	-38.85 dB	
	FNBW	28°	26 °	28 °	
Case-3	FSLs	-19.08 dB	-18.28 dB	-19.15 dB	
	FNs	-38.84 dB	-39.48 dB	-38.16 dB	
	FNBW	28°	26 °	28 °	
Case-4	FSLs	-19.08 dB	-27.73 dB	-19.18 dB	
	FNs	-38.84 dB	-32.65 dB	-36.87 dB	
	FNBW	28°	30 °	28 °	

Table-1. Comparative results

The degree of recovery of each case is further examined and compared using linear correlation coefficient (LCC), median absolute deviation correlation coefficient (MADCC), linear least square regression (LLSR) and root mean square error (RMSE) statistical methods [31]. The results after examination are tabulated in Table-2.

Cases	LCC		MADCC		LLSR		RMSE (%)	
	O-F	O-P	O-F	O-P	O-F	O-P	O-F	O-P
Case-1	0.9965	0.9999	0.8564	0.9991	0.9931	0.9999	1.9029	0.0395
Case-2	0.9966	0.9999	0.8838	0.9990	0.9933	0.9999	1.8526	0.0272
Case-3	0.9821	0.9997	-0.0683	0.9679	0.9646	0.9995	4.2718	0.4674
Case-4	0.9883	0.9992	0.7615	0.9643	0.9768	0.9985	4.1567	1.0870

Table-2. Statistical comparison

The first two methods estimate the correlation coefficients, LLSR calculates the linear regression and RMSE gives the errors before and after recovery. In the table, O-F represents estimation between original and fault patterns, whereas O-P signifies the relation between original and PSO optimized patterns. The values in Table-1 and Table-2 illustrate that the recovery process after faults works satisfactorily.

6. CONCLUSION

The recovery of the AF pattern in case of element failure by perturbing the amplitudes of input excitations is discussed. The FF with set PSO parameters was able to produce close optimization even when 20 % fault elements were present in the array. The numerical results and the statistical data presented for all the cases shows that the developed methodology worked well in obtaining the optimized AF pattern with acceptable values of SLLs, FNs and its depths. In present era when any solid state based active array system is equipped with embedded technology where the algorithm can be stored and processed, the developed methodology can easily be adopted for recovering AF pattern of any large and complex planar phased array. However, the present work shows that the performance of the recovery degrades with the number of fault elements present in an array. Therefore for large number of faults, the replacement of TRMs could be the right choice. In future, AF optimization using phase or combination of phase and amplitude perturbation will be attempted as the extension of the present work.

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